

# APPLICATION MANUAL



## High-Speed Dual Video Amplifier IC TK15450M/L

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# High-Speed Dual Video Amplifier IC TK15450M/L

## 1. DESCRIPTION

The TK15450M/L is a video amplifier IC capable of driving 75 Ω. There are two general-purpose amplifiers in one package.

The TK15450M/L, enclosed in the SOP-8 and SOT23L-8 packages, are the high-speed versions of the TK15420M.

## 2. FEATURES

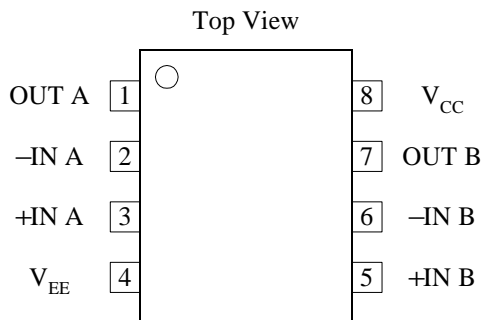
- High Frequency Operation (Flat to 30MHz)
- General-purpose Amplifier Type
- Can Drive Two Video Loads (75Ω)

## 3. APPLICATIONS

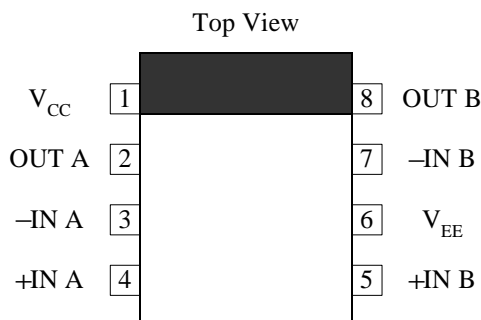
- Video and RF Signal Amplification
- 75Ω Driver

## 4. PIN CONFIGURATION

### ■ SOP-8 (TK15450M)

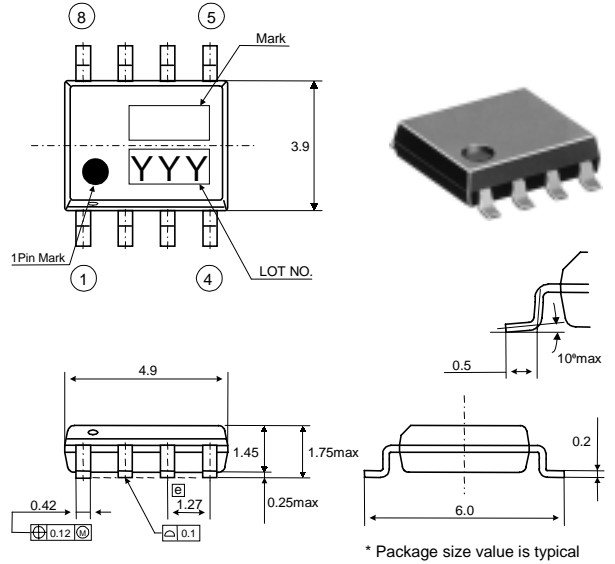


### ■ SOT23L-8 (TK15450L)

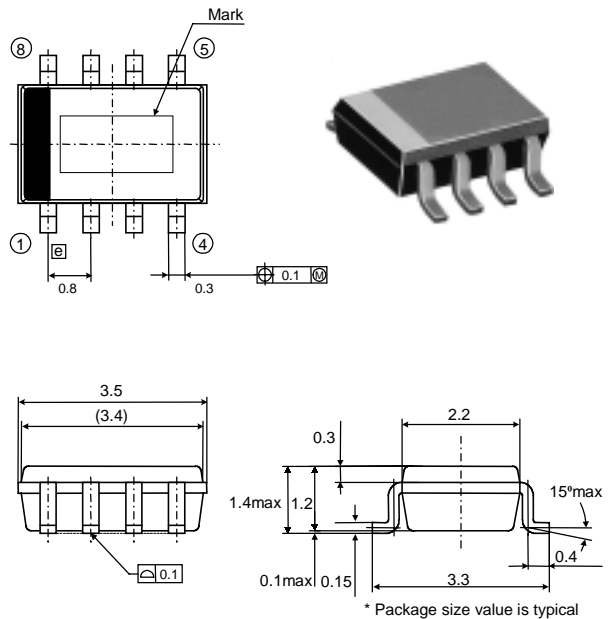


## 5. PACKAGE OUTLINE

### ■ SOP-8 (TK15450M)

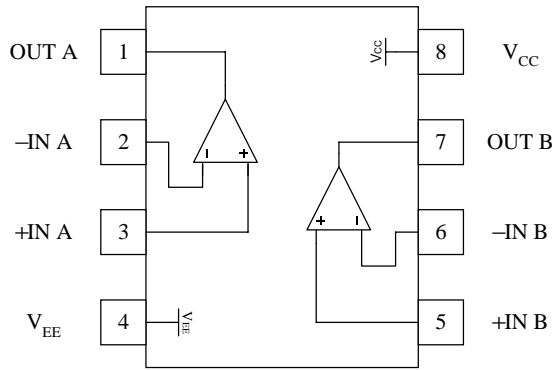


### ■ SOT23L-8 (TK15450L)

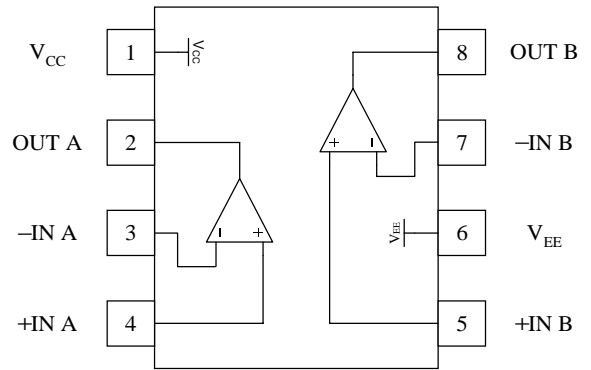


**6. BLOCK DIAGRAM**

■ SOP-8 (TK15450M)



■ SOT23L-8 (TK15450L)



**7. ABSOLUTE MAXIMUM RATINGS**

$T_a=25^{\circ}\text{C}$

Parameter	Symbol	Rating	Units	Conditions
Supply Voltage	$V_{CC}$	12.0	V	
Power Dissipation (SOP-8)	$P_D$	1500	mW	Board Mounted *1
Power Dissipation (SOT23L-8)	$P_D$	1200	mW	Board Mounted *2
Storage Temperature Range	$T_{stg}$	-55 ~ +150	$^{\circ}\text{C}$	
Operating Temperature Range	$T_{OP}$	-25 ~ +75	$^{\circ}\text{C}$	
Maximum operating Frequency	$f_{MAX}$	~ 200	MHz	
Operating Voltage Range	$V_{OP}$	4.0 ~ 11.0	V	

\*1  $P_D$  must be decreased at the rate of 12.0mW/ $^{\circ}\text{C}$  for operation above 25 $^{\circ}\text{C}$ .

\*2  $P_D$  must be decreased at the rate of 9.6mW/ $^{\circ}\text{C}$  for operation above 25 $^{\circ}\text{C}$ .

**8. ELECTRICAL CHARACTERISTICS**

**8-1. ±5V Dual Supply Electrical Characteristics**

$V_{CC}/V_{EE}=\pm 5.0V, T_a=25^{\circ}C, R_L=150\Omega$

Parameter	Symbol	Value			Units	Conditions
		MIN	TYP	MAX		
Supply Current	$I_{CC}$	–	35.3	55.0	mA	No Signal
Output Source Current	$I_{SO}$	–	60	–	mA	
Output Sink Current	$I_{SI}$	–	60	–	mA	
Input Bias Current	$I_{IB}$	–	19.8	60	$\mu A$	
Input Offset Voltage	$V_{IO}$	–	22	–	mV	
Open Circuit Voltage Gain	$G_{VO}$	–	40	–	dB	
Gain-Bandwidth Product	GB	–	200	–	MHz	
Slew Rate	SR	–	320	–	V/ $\mu s$	
Voltage Gain	$G_V$	5.3	5.8	6.3	dB	$f_{in}=1MHz$
Frequency Response	fr	–	0.0	–	dB	$f_{in}=1MHz/30MHz$
Differential Gain	DG	-3	0.2	3	%	$V_{in}=1.0V_{p-p}$ 10 stair case
Differential Phase	DP	-3	0.1	3	deg	$V_{in}=1.0V_{p-p}$ 10 stair case
Cross Talk1	CT1	–	-63	-50	dB	$f_{in}=4.43MHz, V_{in}=1.0V_{p-p}$
Cross Talk2	CT2	–	-45	–	dB	$f_{in}=30.0MHz, V_{in}=1.0V_{p-p}$
Supply Voltage Rejection Ratio1	SVRR1	–	-52	–	dB	$\Delta V_{CC}=0.4V_{p-p}, f=100kHz$
Supply Voltage Rejection Ratio2	SVRR2	–	-41	–	dB	$\Delta V_{EE}=0.4V_{p-p}, f=100kHz$
Input Capacitance	$C_{in}$	–	1.1	–	pF	SOP-8 Package
Input Capacitance	$C_{in}$	–	0.8	–	pF	SOT23L-8 Package

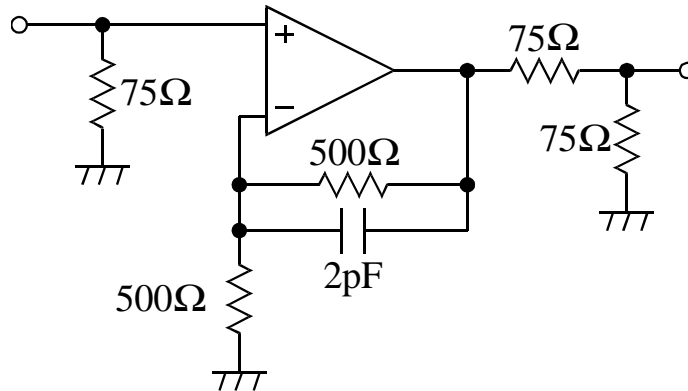
**8-2. 5V Single Supply Electrical Characteristics (Reference Data)**

$V_{CC}=5.0V, T_a=25^{\circ}C, R_L=150\Omega$

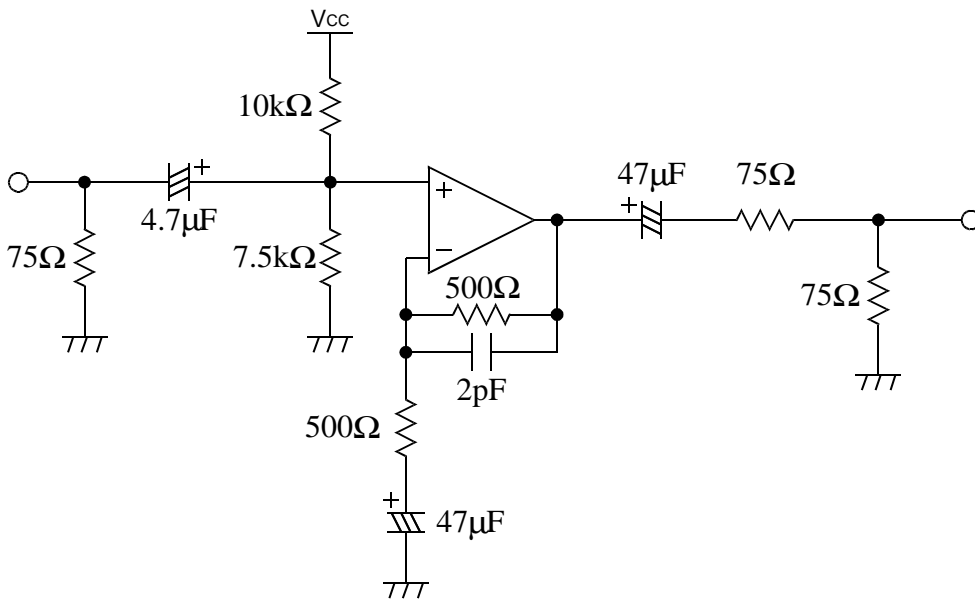
Parameter	Symbol	Value			Units	Conditions
		MIN	TYP	MAX		
Supply Current	$I_{CC}$	–	30.4	–	mA	No Signal
Output Source Current	$I_{SO}$	–	50	–	mA	
Output Sink Current	$I_{SI}$	–	50	–	mA	
Input Bias Current	$I_{IB}$	–	15	–	$\mu A$	
Input Offset Voltage	$V_{IO}$	–	10	–	mV	
Open Circuit Voltage Gain	$G_{VO}$	–	40	–	dB	
Gain-Bandwidth Product	GB	–	200	–	MHz	
Slew Rate	SR	–	180	–	V/ $\mu s$	
Voltage Gain	$G_V$	–	5.8	–	dB	$f_{in}=1MHz$
Frequency Response	fr	–	0.0	–	dB	$f_{in}=1MHz/30MHz$
Differential Gain	DG	–	0.4	–	%	$V_{in}=1.0V_{p-p}$ 10 stair case
Differential Phase	DP	–	0.1	–	deg	$V_{in}=1.0V_{p-p}$ 10 stair case
Cross Talk1	CT1	–	-61	–	dB	$f_{in}=4.43MHz, V_{in}=1.0V_{p-p}$
Cross Talk2	CT2	–	-44	–	dB	$f_{in}=30.0MHz, V_{in}=1.0V_{p-p}$
Supply Voltage Rejection Ratio	SVRR	–	-44	–	dB	$\Delta V_{CC}=0.4V_{p-p}, f=100kHz$

**9. TEST CIRCUIT**

**9-1. ±5V Dual Supply Test Circuit**

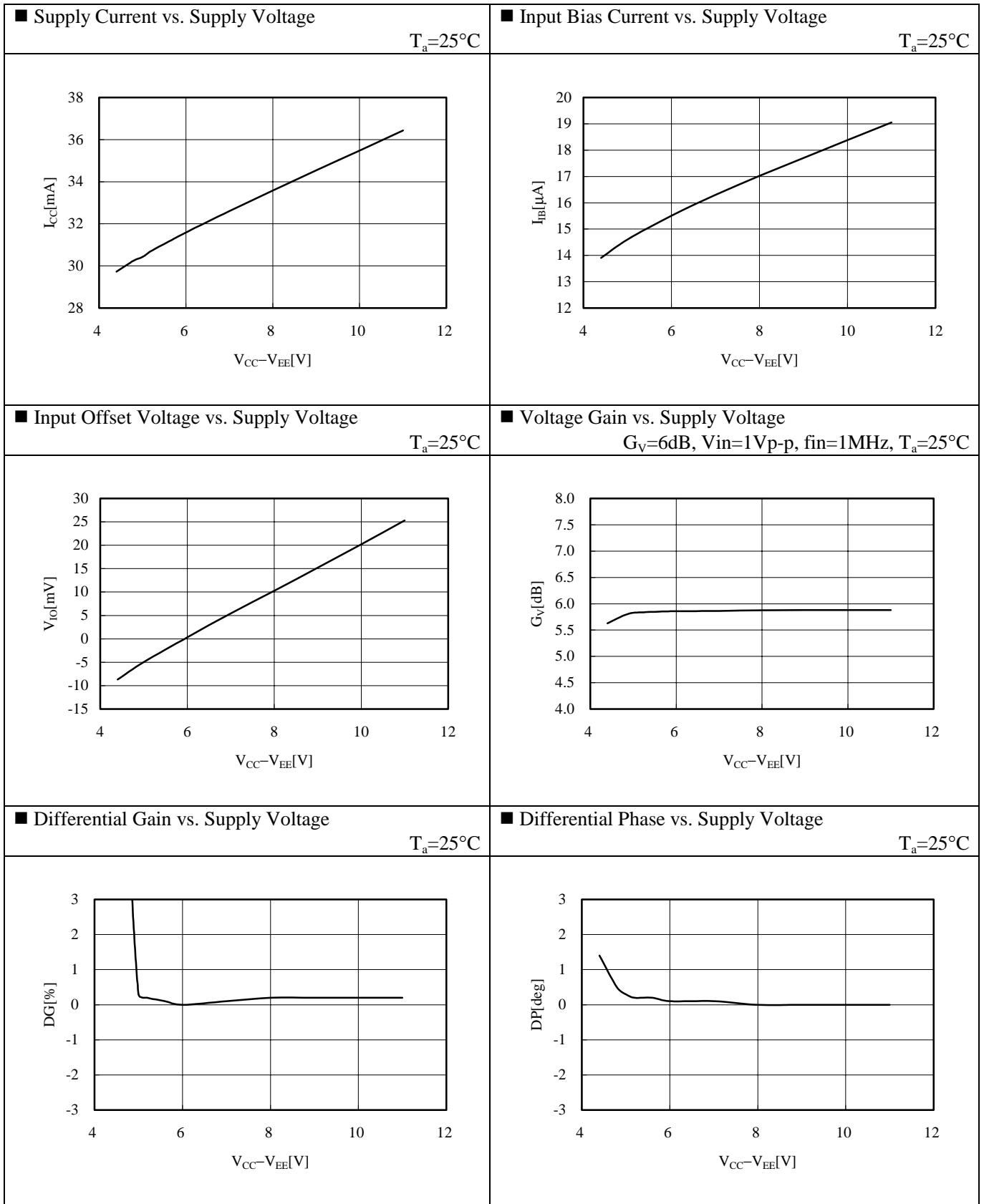


**9-2. 5V Single Supply Test Circuit**

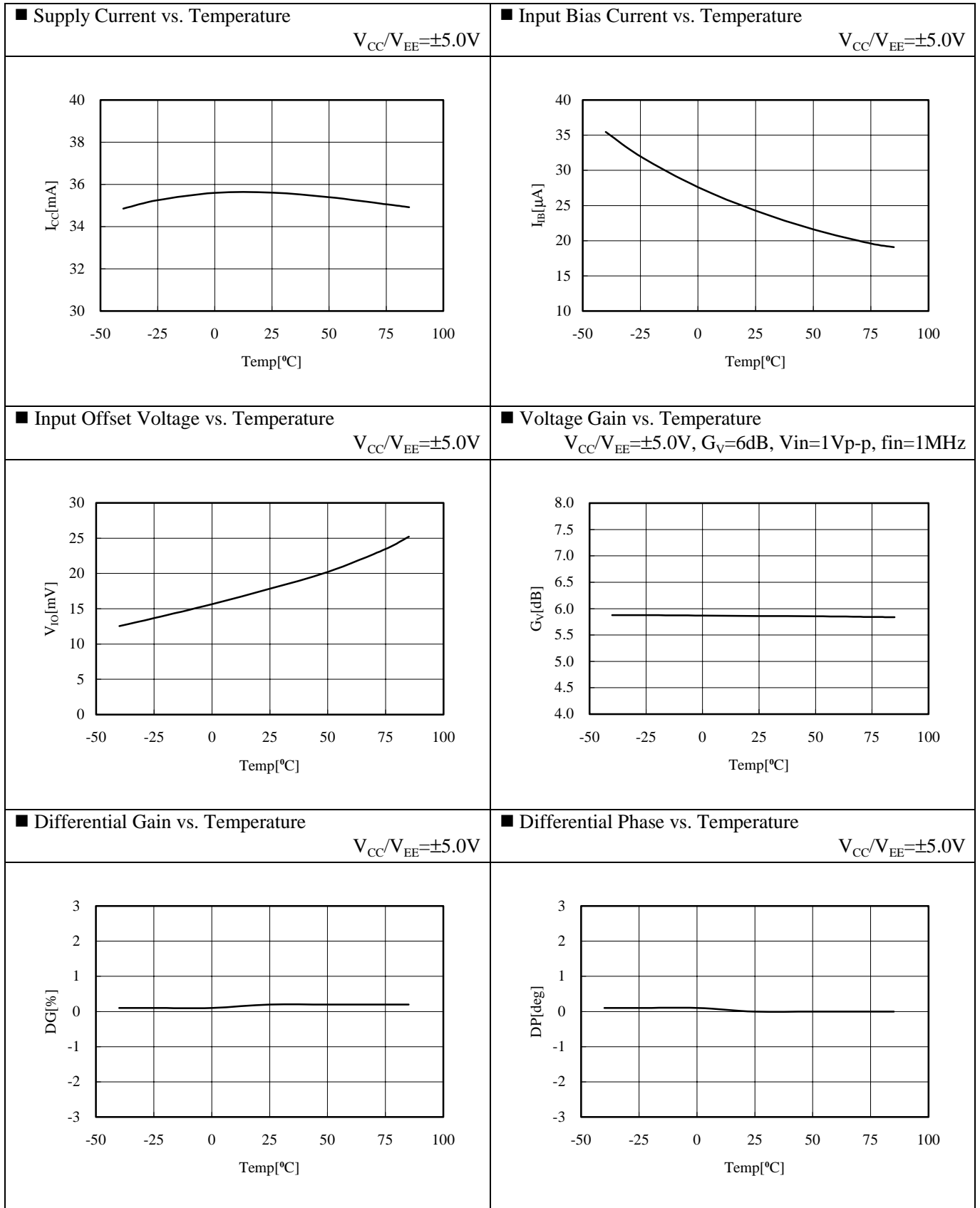


**10. TYPICAL CHARACTERISTICS**

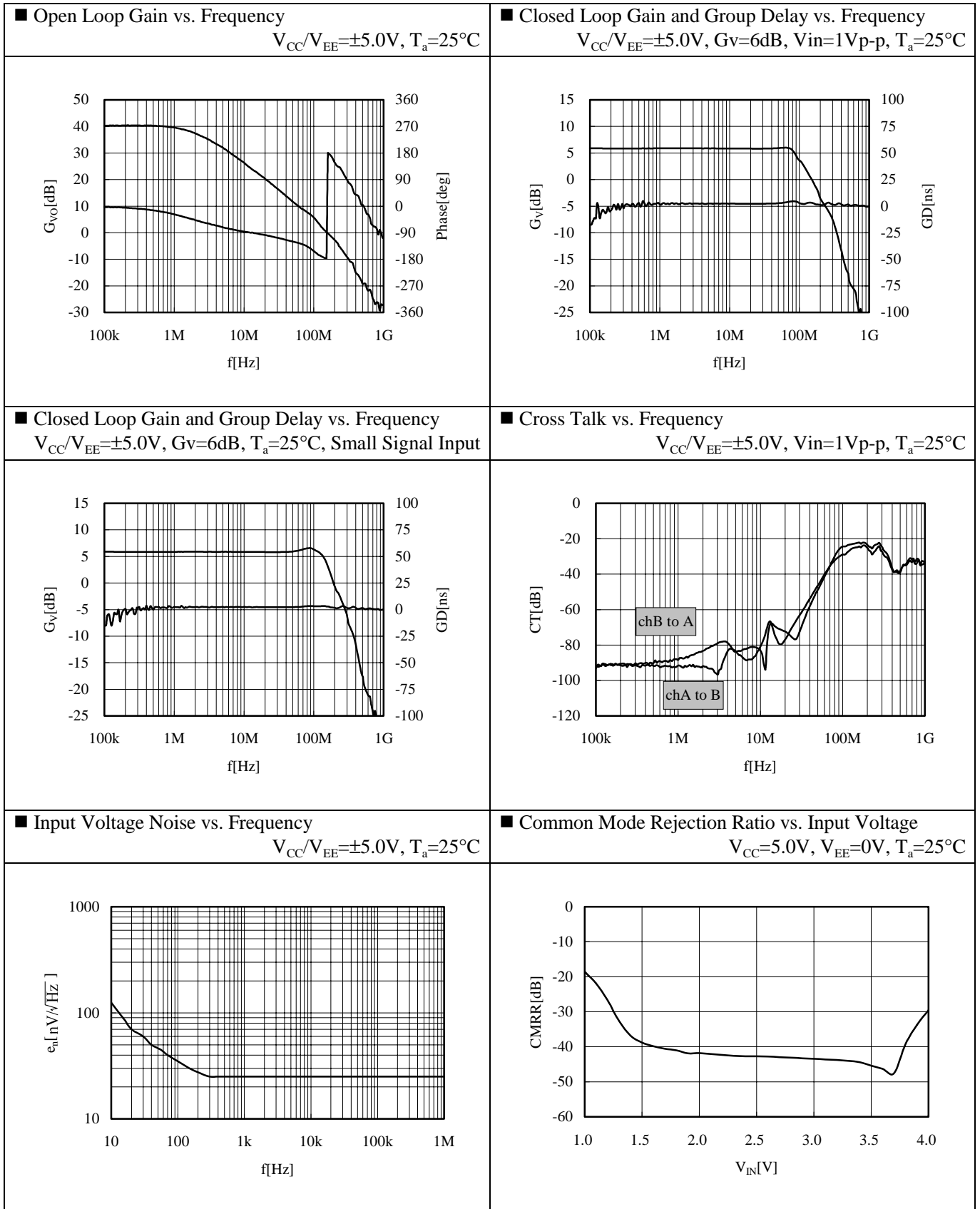
**10-1. Supply Voltage Characteristics**



**10-2. Temperature Characteristics**

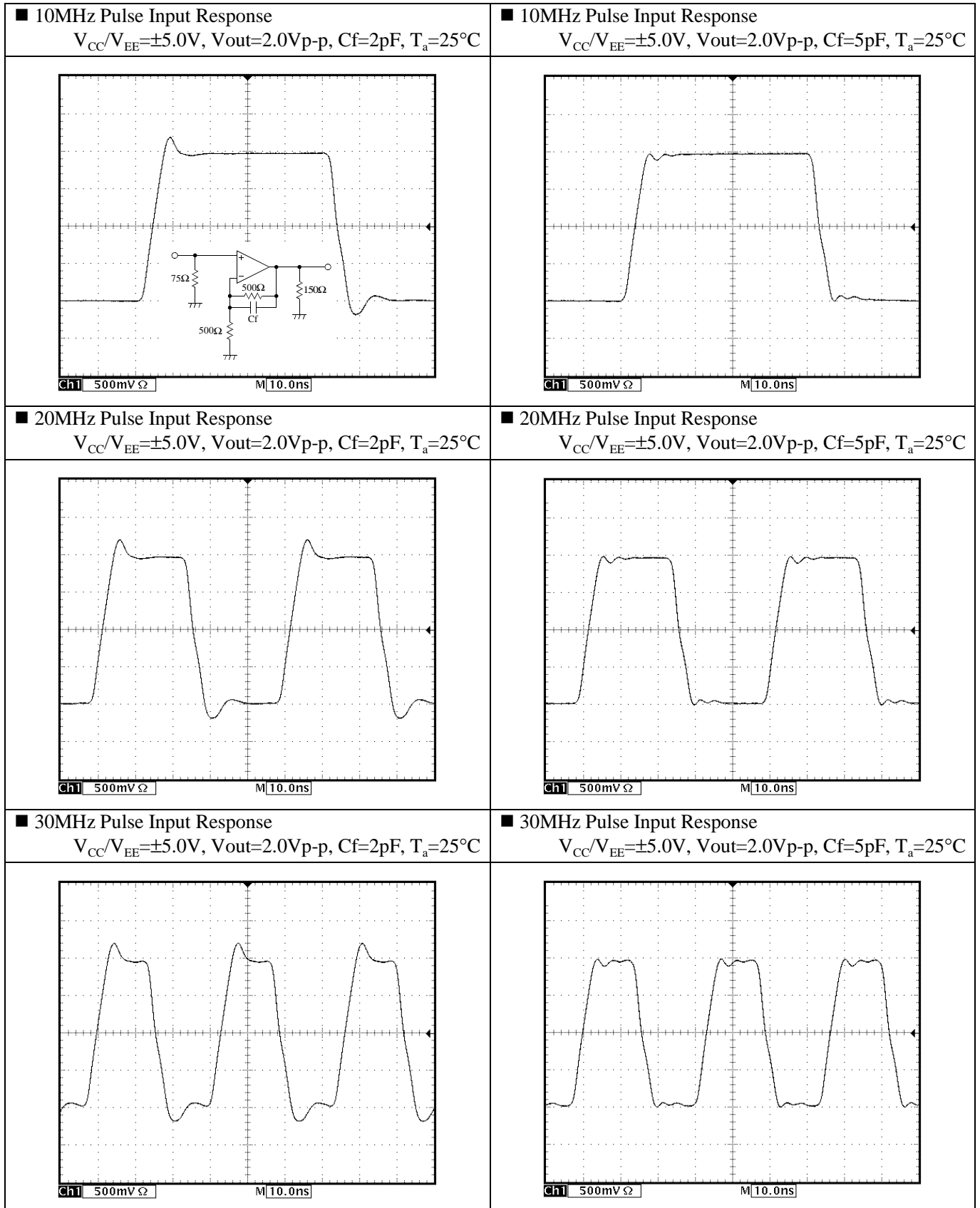


**10-3. Various Main Characteristics**

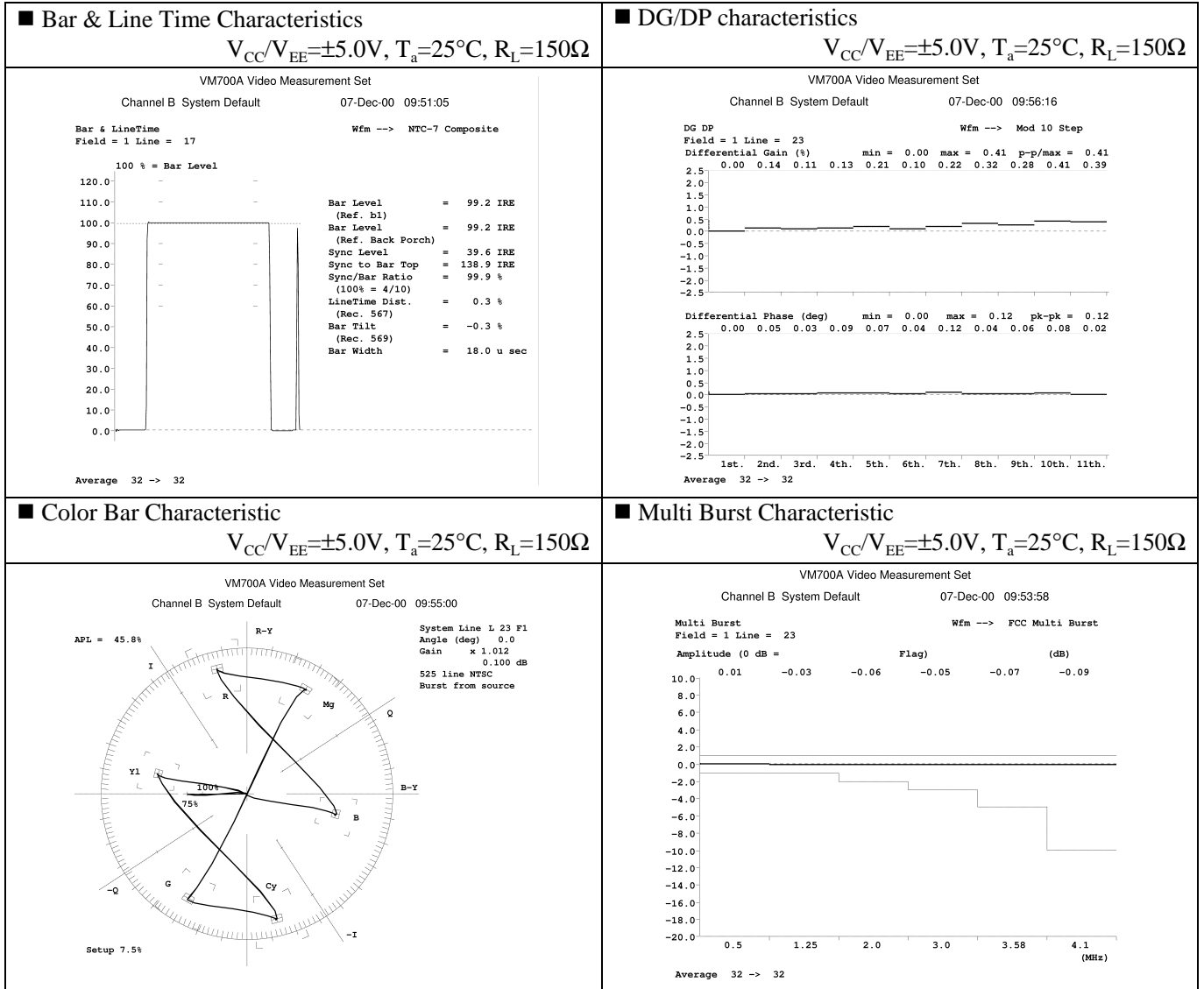




**10-4. Pulse Input Characteristics**



**10-5. Video Signal Characteristics**



**11. PIN DESCRIPTION**

SOP-8 Package

Pin No.	Name	Internal Equivalent Circuit	Description
1	OUT A		Output terminal of channel A.
2 3	-IN A +IN A		Inverting and Non-inverting input terminals of channel A. This circuit is a differential amplifier structure using NPN transistors.
4	V <sub>EE</sub>		Negative Power Supply Terminal.
5 6	+IN B -IN B		Inverting and Non-inverting input terminals of channel B. This circuit is a differential amplifier structure using NPN transistors.
7	OUT B		Output terminal of channel A.
8	V <sub>CC</sub>		Positive Power Supply Terminal.

12. APPLICATIONS INFORMATION

12-1. About Amplitude Restrictions

In certain applications, the output voltage is limited by the input voltage. This is explained in the outline below using the internal equivalent circuit shown in Figure 1.

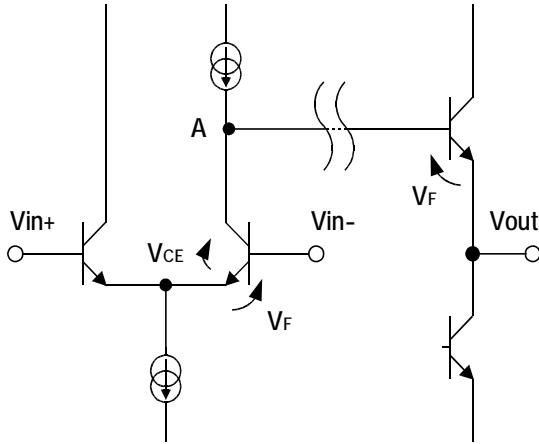


Figure 1: The internal equivalent circuit

From Figure 1, if the voltage  $V_A$  at A point is shown from the input side and the output side respectively, the expression is as follows:

$$V_A \geq V_{in} - V_F + V_{CE} \tag{1}$$

$$V_A = V_{out} + V_F \tag{2}$$

Thus

$$V_{out} - V_{in} + 2V_F \geq V_{CE} \tag{3}$$

Substitution of  $V_F = 0.7V$  into (3) gives

$$V_{out} - V_{in} + 1.4V \geq V_{CE} \tag{4}$$

Depending on the relationship between  $V_{out}$  and  $V_{in}$ , it may become impossible to secure the Saturation voltage  $V_{CE}$  (about 0.3 V) of the inverting input transistor; as a result, the linearity of the input and output voltage will collapse.

An example of this application is shown in Figure 2 with the measures explained below.

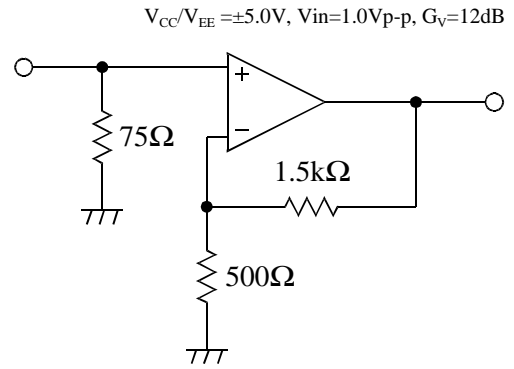


Figure 2: Application example

In Figure 2, if -0.5V (the minimum value of input amplitude) is given to the input, the output voltage will be set to -2.0V. Substitution of  $V_{in}$  and  $V_{out}$  into (4) gives

$$V_{out} - V_{in} + 1.4V = -0.1V \leq V_{CE} (0.3V) \tag{5}$$

This shows that the transistor of the inverting input is operating in the saturation region; for this reason, it becomes impossible to keep linearity of the input-to-output voltage. As shown in Figure 3, there is a method of providing  $V_{REF}$  as a preventive measure. It is possible to raise the output voltage by setting up  $V_{REF}$  appropriately, and avoid amplitude restrictions.

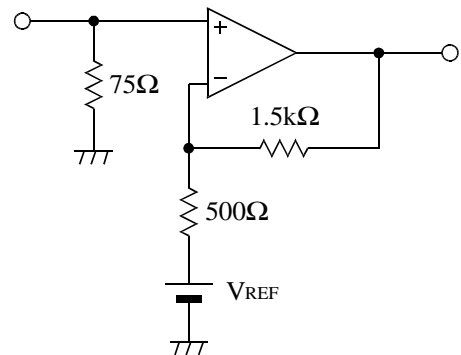


Figure 3: Example of preventive measures

If the input voltage and  $V_{REF}$  are assumed to be -0.5V, the output voltage also becomes -0.5V.

This result is substituted into expression (4)

$$V_{out} - V_{in} + 1.4V = 1.4V \geq V_{CE} (0.3V) \tag{6}$$

As a result, the saturation voltage of the inverting input transistor is secured, and the amplitude limitation can be avoided. However, it is necessary to pay attention to the dynamic range, especially when using this IC with a low power supply. This method may be used to control the output bias voltage.

### 12-2. Two-Line Video Driver

As shown in Figure 4, this IC is a useful circuit for driving two-video lines.

Figure 5 and Figure 6 show the frequency characteristic and the DG/DP characteristics.

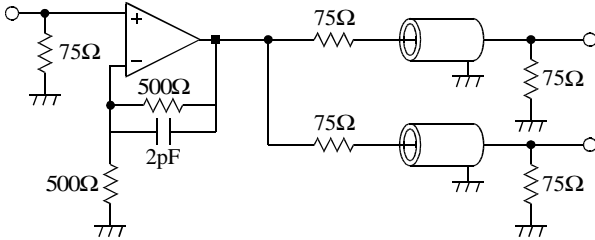


Figure 4: Two-line video driver

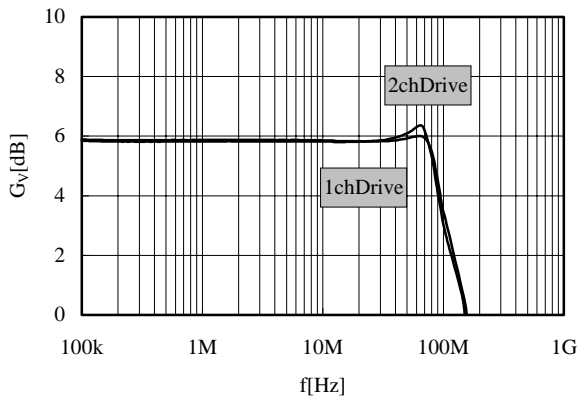


Figure 5: Frequency characteristics

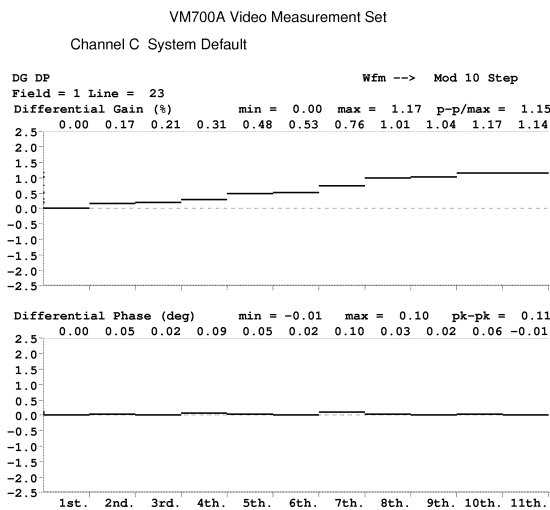


Figure 6: DG/DP characteristics

### 12-3. About Frequency Characteristic

If the gain is raised, the frequency characteristic falls on the balance with the open loop characteristic; this may end realizing the 30MHz flat characteristic. In this case, the frequency characteristic can be improved by a combination application shown in Figure 7.

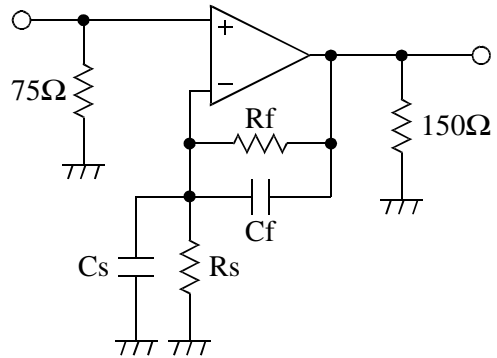


Figure 7: Example of designing a 30MHz flat circuit

The frequency characteristics shown in Figure 8 can be obtained by setting each constant according to the gain as follows.

- $G_v=6\text{dB}$       $R_f=500\Omega, R_s=470\Omega, C_f=2\text{pF}$
- $G_v=9\text{dB}$       $R_f=470\Omega, R_s=240\Omega$
- $G_v=12\text{dB}$      $R_f=470\Omega, R_s=150\Omega, C_s=6\text{pF}$

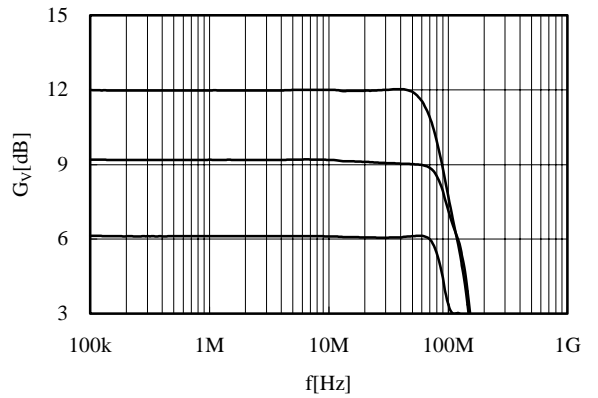


Figure 8: 30MHz flatness characteristics

### 12-4. Driving Capacitive Loads

If a capacitive load ( $C_L$ ) is assumed in the measurement circuit of Figure 9, and the gain is measured, the resulting frequency characteristic is shown in Figure 10.

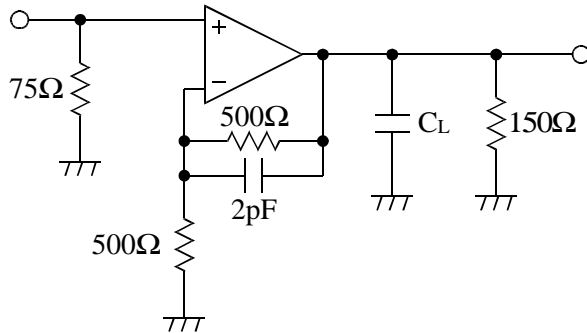


Figure 9: Measurement circuit

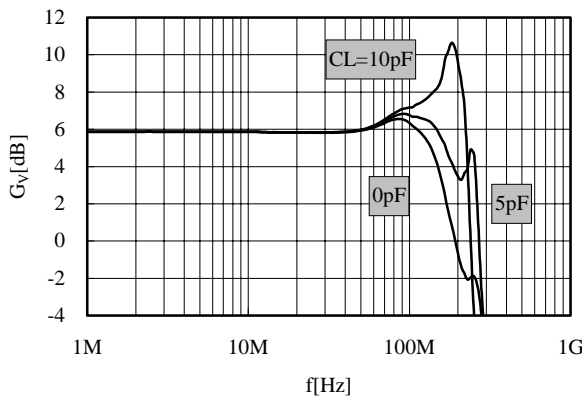


Figure 10: Frequency characteristics of capacitive load

Therefore, please note that oscillation may occur when a capacitive load is directly connected to the output terminal.

### 12-5. Use as a Buffer Amplifier

The gain of this operational amplifier IC can be changed with the external parts.

However, if this IC is used as a buffer amplifier (Figure 11), peaking is generated as shown in Figure 12.

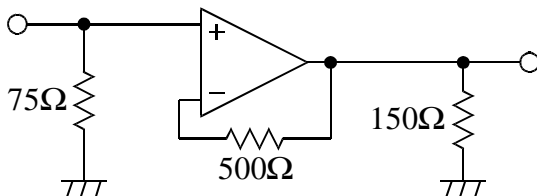


Figure 11: Example of using buffer amplifier

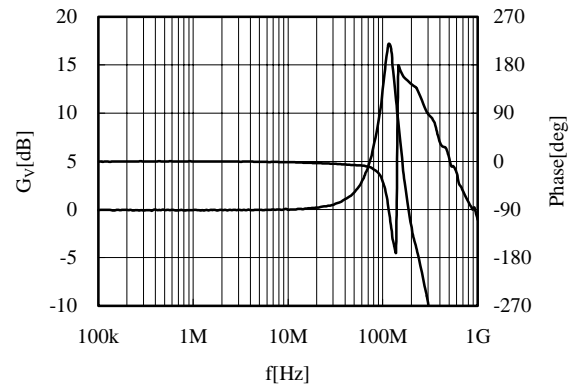


Figure 12: Frequency characteristic of buffer amplifier

There is a possibility of oscillation, so please do not use the amplifier as a buffer.

### 12-6. Differential Input Signal Swing

Application of this IC as a differential Amplifier is also possible. One example is shown in Figure 13, with the waveforms of each terminal shown in Figure 14.

The use of 5.0V single power supply is assumed.

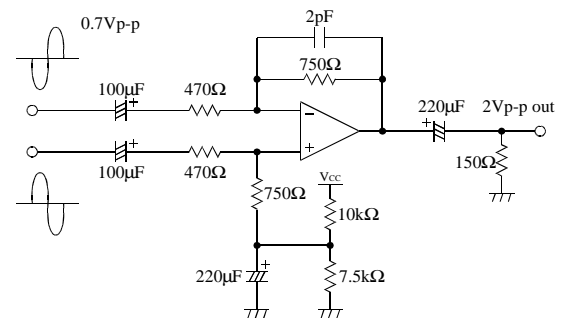


Figure 13: Differential input signal application

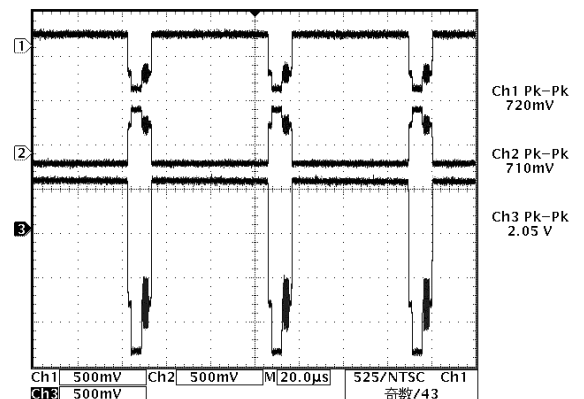


Figure 14: Waveforms of each terminal

Waveforms are non-inverting input, inverting input and output from the top.

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