

5 V, SILICON GERMANIUM MMIC MEDIUM OUTPUT POWER AMPLIFIER

DESCRIPTION

The μ PC3232TB is a silicon germanium (SiGe) monolithic integrated circuit designed as IF amplifier for DBS tuners. This IC is manufactured using our 50 GHz f_{\max} UHS2 (Ultra High Speed Process) SiGe bipolar process.

FEATURES

- Low current : $I_{CC} = 26.0$ mA TYP.
- Medium output power : $P_{O(sat)} = +15.5$ dBm TYP. @ $f = 1.0$ GHz
: $P_{O(sat)} = +12.0$ dBm TYP. @ $f = 2.2$ GHz
- High linearity : $P_{O(1\text{ dB})} = +11.0$ dBm TYP. @ $f = 1.0$ GHz
: $P_{O(1\text{ dB})} = +8.5$ dBm TYP. @ $f = 2.2$ GHz
- Power gain : $G_P = 32.8$ dB MIN. @ $f = 1.0$ GHz
: $G_P = 33.5$ dB MIN. @ $f = 2.2$ GHz
- Gain flatness : $\Delta G_P = 1.0$ dB TYP. @ $f = 1.0$ to 2.2 GHz
- Noise figure : $NF = 4$ dB TYP. @ $f = 1.0$ GHz
: $NF = 4.1$ dB TYP. @ $f = 2.2$ GHz
- Supply voltage : $V_{CC} = 4.5$ to 5.5 V
- Port impedance : input/output 50 Ω

APPLICATIONS

- IF amplifiers in LNB for DBS converters etc.

ORDERING INFORMATION

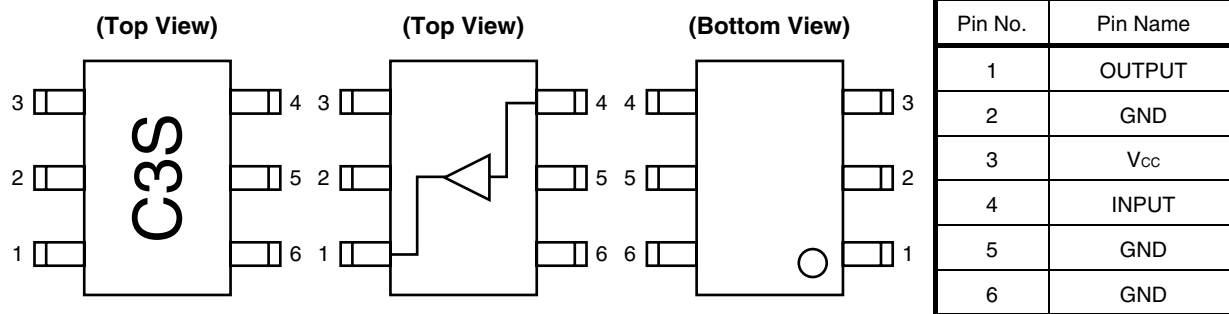
| Part Number | Order Number | Package | Marking | Supplying Form |
|-------------------|---------------------|--------------------------------|---------|---|
| μ PC3232TB-E3 | μ PC3232TB-E3-A | 6-pin super minimold (Pb-Free) | C3S | <ul style="list-style-type: none"> • Embossed tape 8 mm wide • Pin 1, 2, 3 face the perforation side of the tape • Qty 3 kpcs/reel |

Remark To order evaluation samples, please contact your nearby sales office
Part number for sample order: μ PC3232TB

Caution Observe precautions when handling because these devices are sensitive to electrostatic discharge.

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Not all products and/or types are available in every country. Please check with an NEC Electronics sales representative for availability and additional information.

PIN CONNECTIONS



PRODUCT LINE-UP OF 5 V-BIAS SILICON MMIC MEDIUM OUTPUT POWER AMPLIFIER
 (T_A = +25°C, f = 1 GHz, V_{CC} = V_{out} = 5.0 V, Z_S = Z_L = 50 Ω)

| Part No. | P _{O (sat)} (dBm) | G _P (dB) | NF (dB) | I _{CC} (mA) | Package | Marking |
|-----------|-------------------------------|------------------------|---------------------|-------------------------|----------------------|---------|
| μPC2708TB | +10.0 | 15.0 | 6.5 | 26 | 6-pin super minimold | C1D |
| μPC2709TB | +11.5 | 23.0 | 5.0 | 25 | | C1E |
| μPC2710TB | +13.5 | 33.0 | 3.5 | 22 | | C1F |
| μPC2776TB | +8.5 | 23.0 | 6.0 | 25 | | C2L |
| μPC3223TB | +12.0 | 23.0 | 4.5 | 19 | | C3J |
| μPC3225TB | +15.5 ^{Note} | 32.5 ^{Note} | 3.7 ^{Note} | 24.5 | | C3M |
| μPC3226TB | +13.0 | 25.0 | 5.3 | 15.5 | | C3N |
| μPC3232TB | +15.5 | 32.8 | 4.0 | 26 | | C3S |

Note μPC3225TB is f = 0.95 GHz

Remark Typical performance. Please refer to **ELECTRICAL CHARACTERISTICS** in detail.

ABSOLUTE MAXIMUM RATINGS

| Parameter | Symbol | Conditions | Ratings | Unit |
|-------------------------------|------------------|------------------------------------|-------------|------|
| Supply Voltage | V _{CC} | T _A = +25°C | 6.0 | V |
| Total Circuit Current | I _{CC} | T _A = +25°C | 45 | mA |
| Power Dissipation | P _D | T _A = +85°C Note | 270 | mW |
| Operating Ambient Temperature | T _A | | -40 to +85 | °C |
| Storage Temperature | T _{stg} | | -55 to +150 | °C |
| Input Power | P _{in} | T _A = +25°C | 0 | dBm |

Note Mounted on double-sided copper-clad 50 × 50 × 1.6 mm epoxy glass PWB

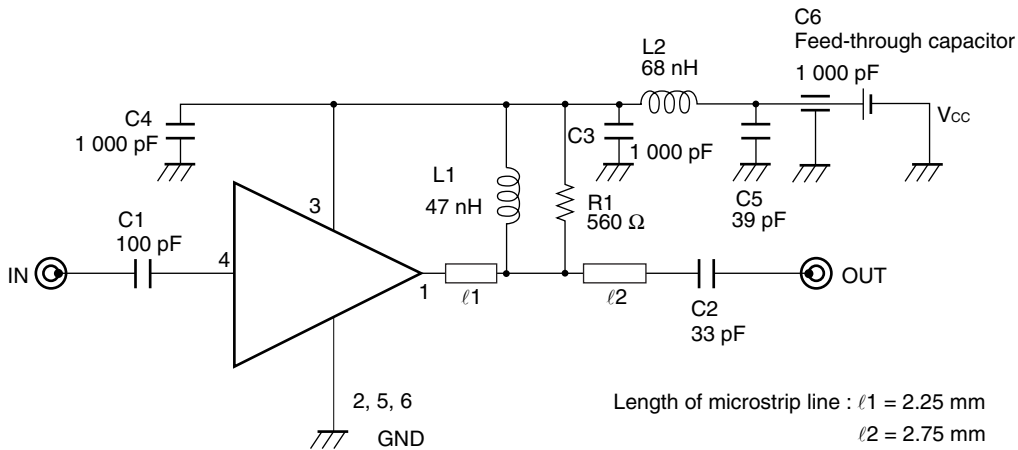
RECOMMENDED OPERATING RANGE

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
|-------------------------------|-----------------|------------|------|------|------|------|
| Supply Voltage | V _{CC} | | 4.5 | 5.0 | 5.5 | V |
| Operating Ambient Temperature | T _A | | -40 | +25 | +85 | °C |

ELECTRICAL CHARACTERISTICS (T_A = +25°C, V_{CC} = V_{out} = 5.0 V, Z_S = Z_L = 50 Ω)

| Parameter | Symbol | Test Conditions | MIN. | TYP. | MAX. | Unit |
|---|-------------------------|---|------|-------|------|------|
| Circuit Current | I _{CC} | No input signal | 20 | 26 | 32 | mA |
| Power Gain 1 | G _{P1} | f = 0.25 GHz, P _{in} = -35 dBm | 29 | 31.5 | 34 | dB |
| Power Gain 2 | G _{P2} | f = 1.0 GHz, P _{in} = -35 dBm | 30 | 32.8 | 35.5 | |
| Power Gain 3 | G _{P3} | f = 1.8 GHz, P _{in} = -35 dBm | 31 | 33.8 | 37 | |
| Power Gain 4 | G _{P4} | f = 2.2 GHz, P _{in} = -35 dBm | 30.5 | 33.5 | 36.5 | |
| Power Gain 5 | G _{P5} | f = 2.6 GHz, P _{in} = -35 dBm | 29 | 32.2 | 35.5 | |
| Power Gain 6 | G _{P6} | f = 3.0 GHz, P _{in} = -35 dBm | 27 | 30.7 | 34 | |
| Gain Flatness | ΔG _P | f = 1.0 to 2.2 GHz, P _{in} = -35 dBm | - | 1.0 | - | dB |
| K factor 1 | K1 | f = 1.0 GHz, P _{in} = -35 dBm | - | 1.3 | - | - |
| K factor 2 | K2 | f = 2.2 GHz, P _{in} = -35 dBm | - | 1.9 | - | - |
| Saturated Output Power 1 | P _{O (sat) 1} | f = 1.0 GHz, P _{in} = 0 dBm | +13 | +15.5 | - | dBm |
| Saturated Output Power 2 | P _{O (sat) 2} | f = 2.2 GHz, P _{in} = -5 dBm | +9.5 | +12 | - | |
| Gain 1 dB Compression Output Power 1 | P _{O (1 dB) 1} | f = 1.0 GHz | +8 | +11 | - | dBm |
| Gain 1 dB Compression Output Power 2 | P _{O (1 dB) 2} | f = 2.2 GHz | +6 | +8.5 | - | |
| Noise Figure 1 | NF1 | f = 1.0 GHz | - | 4 | 4.8 | dB |
| Noise Figure 2 | NF2 | f = 2.2 GHz | - | 4.1 | 4.9 | |
| Isolation 1 | ISL1 | f = 1.0 GHz, P _{in} = -35 dBm | 36 | 41 | - | dB |
| Isolation 2 | ISL2 | f = 2.2 GHz, P _{in} = -35 dBm | 38 | 45 | - | |
| Input Return Loss 1 | RL _{in1} | f = 1.0 GHz, P _{in} = -35 dBm | 9.5 | 13 | - | dB |
| Input Return Loss 2 | RL _{in2} | f = 2.2 GHz, P _{in} = -35 dBm | 10 | 14.5 | - | |
| Output Return Loss 1 | RL _{out1} | f = 1.0 GHz, P _{in} = -35 dBm | 12 | 15.5 | - | dB |
| Output Return Loss 2 | RL _{out2} | f = 2.2 GHz, P _{in} = -35 dBm | 12 | 15 | - | |
| Input 3rd Order Distortion Intercept Point 1 | IIP ₃₁ | f1 = 1 000 MHz, f2 = 1 001 MHz | - | -9 | - | dBm |
| Input 3rd Order Distortion Intercept Point 2 | IIP ₃₂ | f1 = 2 200 MHz, f2 = 2 201 MHz | - | -15.5 | - | |
| Output 3rd Order Distortion Intercept Point 1 | OIP ₃₁ | f1 = 1 000 MHz, f2 = 1 001 MHz | - | +23.5 | - | dBm |
| Output 3rd Order Distortion Intercept Point 2 | OIP ₃₂ | f1 = 2 200 MHz, f2 = 2 201 MHz | - | +18 | - | |
| 2nd Order Intermodulation Distortion | IM ₂ | f1 = 1 000 MHz, f2 = 1 001 MHz, P _{out} = -5 dBm/tone | - | 50 | - | dBc |
| 2nd Harmonic | 2f ₀ | f ₀ = 1.0 GHz, P _{out} = -15 dBm | - | 70 | - | dBc |

TEST CIRCUIT



The application circuits and their parameters are for reference only and are not intended for use in actual design-ins.

COMPONENTS OF TEST CIRCUIT FOR MEASURING ELECTRICAL CHARACTERISTICS

| | Type | Value |
|--------|------------------------|----------|
| R1 | Chip Resistance | 560 Ω |
| L1 | Chip Inductor | 47 nH |
| L2 | Chip Inductor | 68 nH |
| C1 | Chip Capacitor | 100 pF |
| C2 | Chip Capacitor | 33 pF |
| C3, C4 | Chip Capacitor | 1 000 pF |
| C5 | Chip Capacitor | 39 pF |
| C6 | Feed-through Capacitor | 1 000 pF |

INDUCTOR FOR THE OUTPUT PIN

The internal output transistor of this IC, to output medium power. To supply current for output transistor, connect an inductor between the Vcc pin (pin 3) and output pin (pin 1). Select inductance, as the value listed above.

The inductor has both DC and AC effects. In terms of DC, the inductor biases the output transistor with minimum voltage drop to output enable high level. In terms of AC, the inductor makes output-port impedance higher to get enough gain. In this case, large inductance and Q is suitable (Refer to the following page).

CAPACITORS FOR THE Vcc, INPUT AND OUTPUT PINS

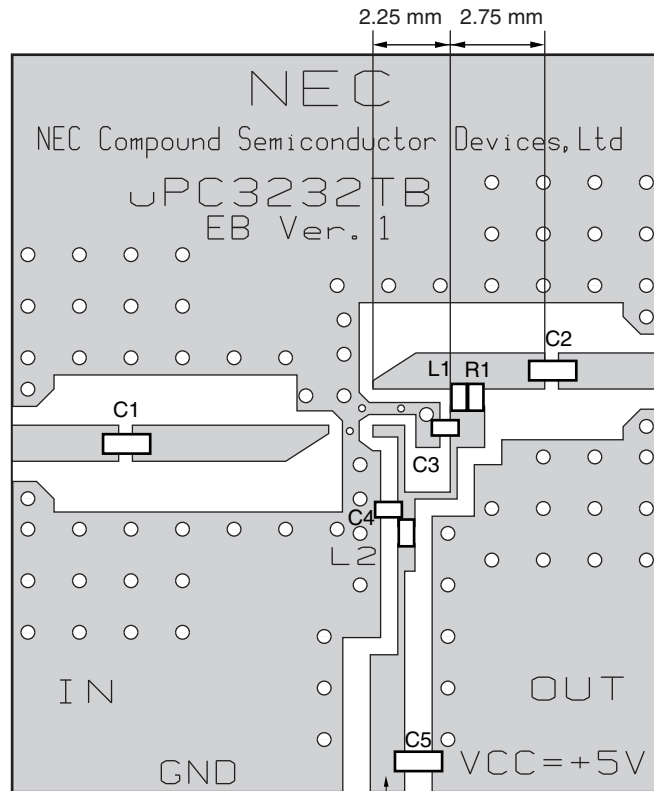
Capacitors of 1 000 pF are recommendable as the bypass capacitor for the Vcc pin and the coupling capacitors for the input and output pins.

The bypass capacitor connected to the Vcc pin is used to minimize ground impedance of Vcc pin. So, stable bias can be supplied against Vcc fluctuation.

The coupling capacitors, connected to the input and output pins, are used to cut the DC and minimize RF serial impedance. Their capacitances are therefore selected as lower impedance against a 50 Ω load. The capacitors thus perform as high pass filters, suppressing low frequencies to DC.

To obtain a flat gain from 100 MHz upwards, 1 000 pF capacitors are used in the test circuit. In the case of under 10 MHz operation, increase the value of coupling capacitor such as 10 000 pF. Because the coupling capacitors are determined by equation, $C = 1/(2 \pi Rf_c)$.

ILLUSTRATION OF THE TEST CIRCUIT ASSEMBLED ON EVALUATION BOARD



C6: Feed-through Capacitor

COMPONENT LIST

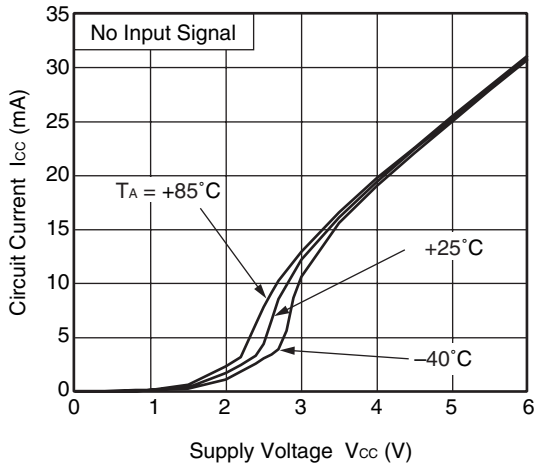
| | Value | Size |
|--------|--------------|------------------------|
| R1 | 560 Ω | 1005 |
| L1 | 47 nH | 1005 |
| L2 | 68 nH | 1005 |
| C1 | 100 pF | 1608 |
| C2 | 33 pF | 1608 |
| C3, C4 | 1 000 pF | 1005 |
| C5 | 39 pF | 1608 |
| C6 | 1 000 pF | Feed-through Capacitor |

Notes

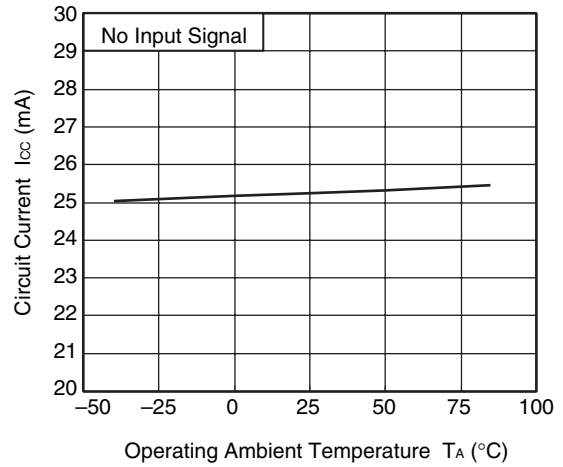
1. 19 × 21.46 × 0.51 mm double sided copper clad RO4003C (Rogers) board.
2. Back side: GND pattern
3. Au plated on pattern
4. \circ : Through holes
5. L1, L2: FDk's products

TYPICAL CHARACTERISTICS ($T_A = +25^\circ\text{C}$, $V_{CC} = 5.0\text{ V}$, $Z_S = Z_L = 50\ \Omega$, unless otherwise specified)

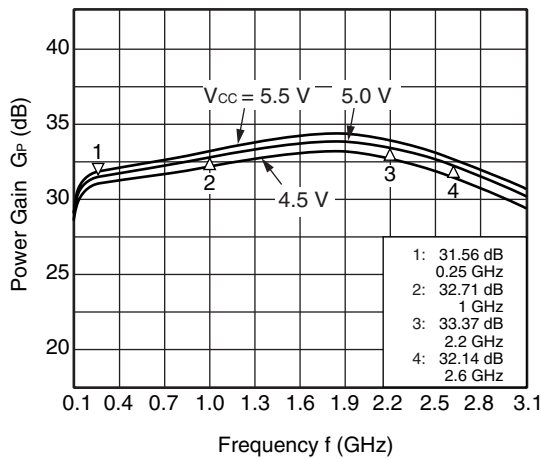
CIRCUIT CURRENT vs. SUPPLY VOLTAGE



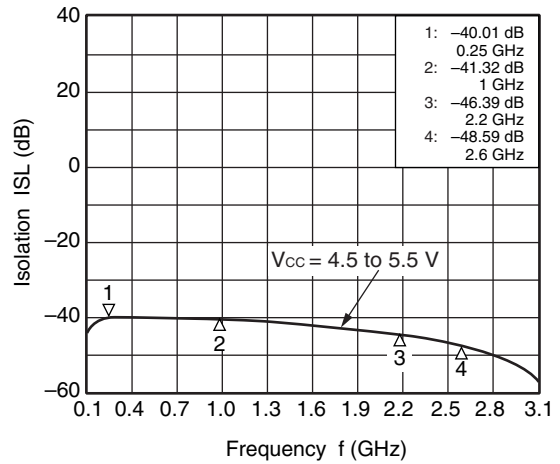
CURCUIT CURRENT vs. OPERATING AMBIENT TEMPERATURE



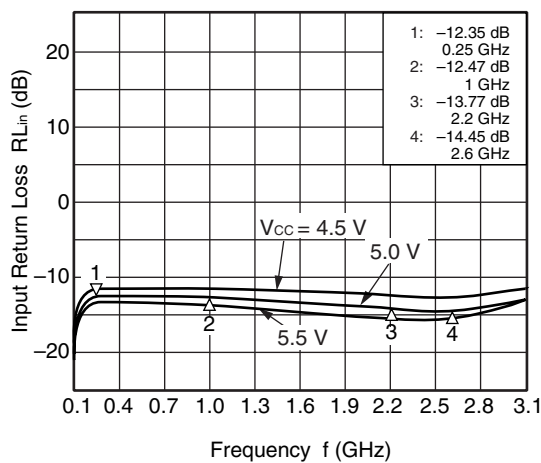
POWER GAIN vs. FREQUENCY



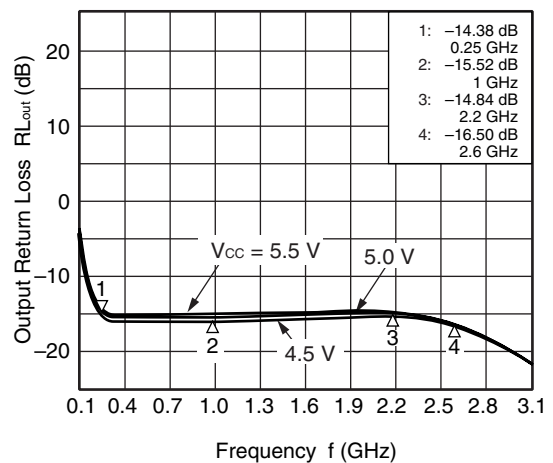
ISOLATION vs. FREQUENCY



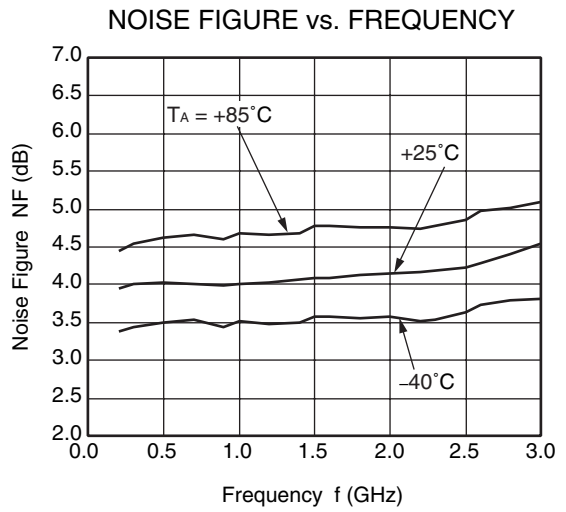
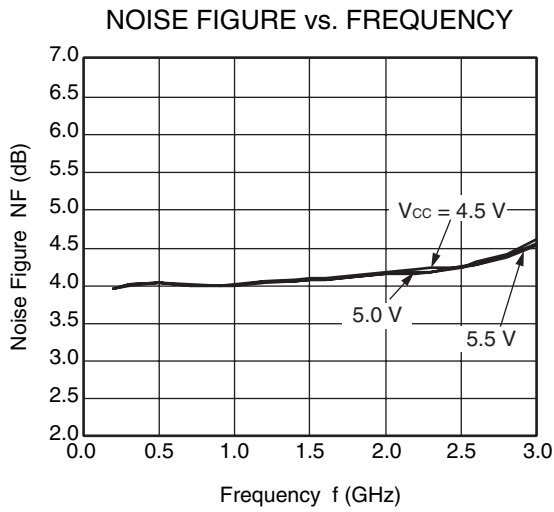
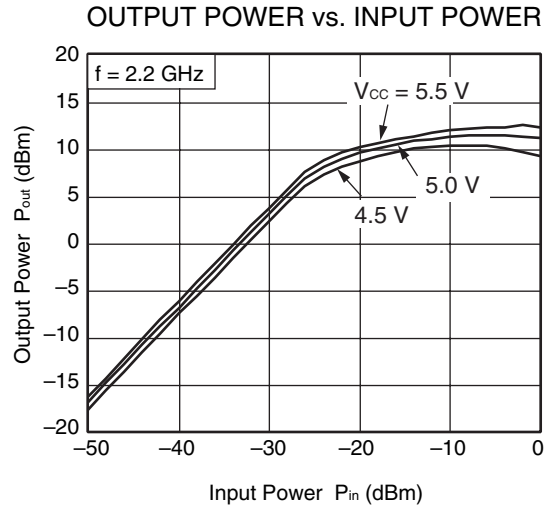
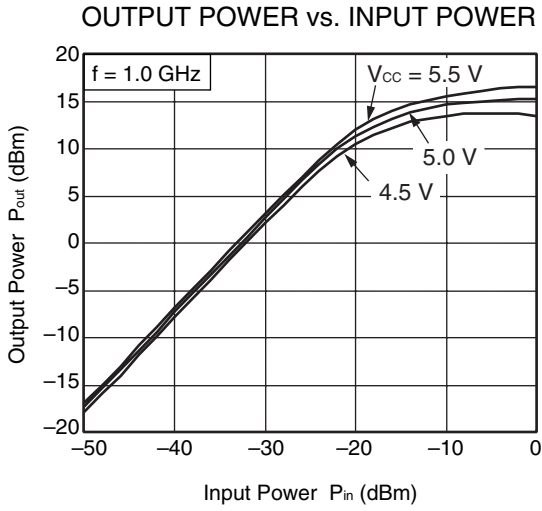
INPUT RETURN LOSS vs. FREQUENCY



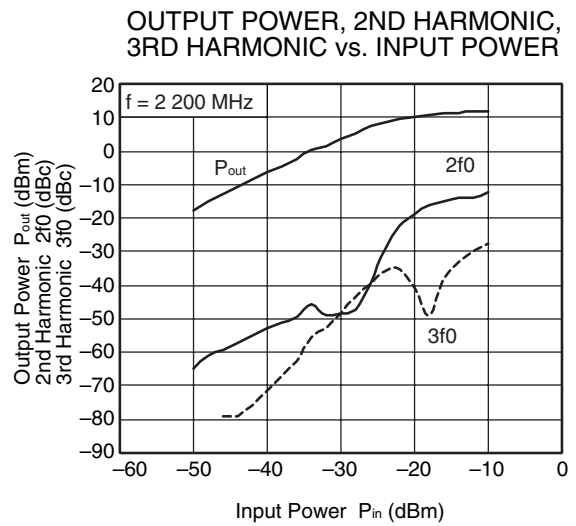
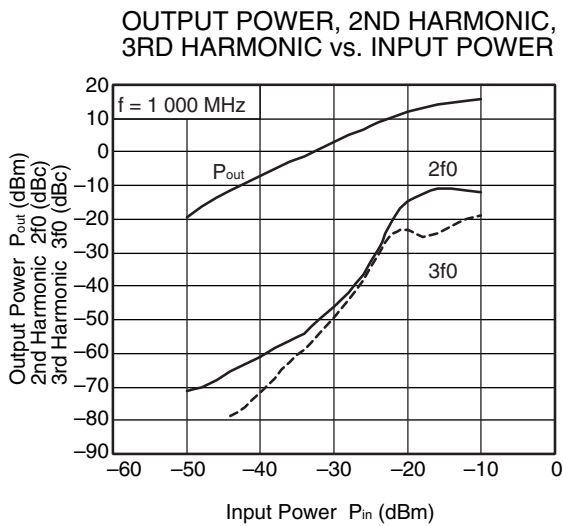
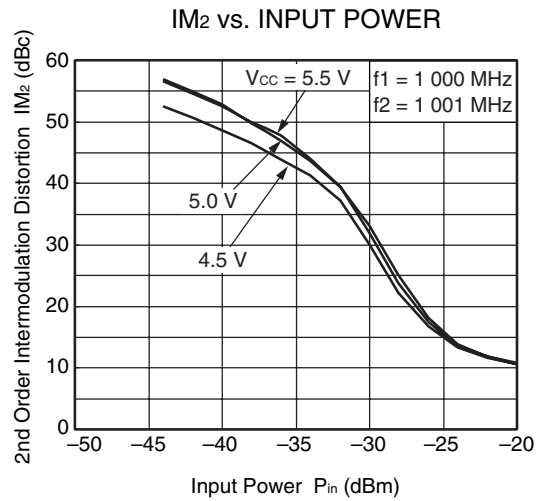
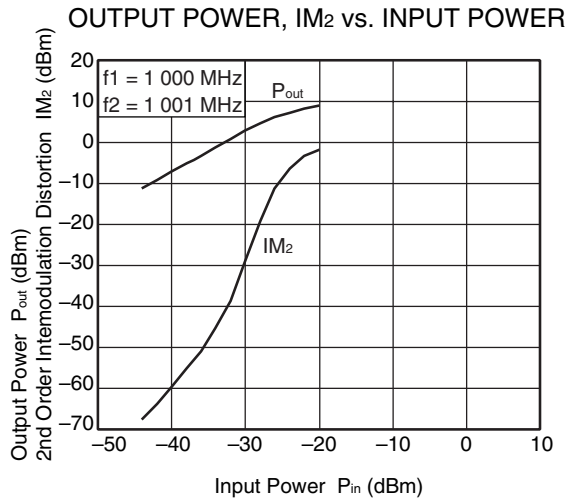
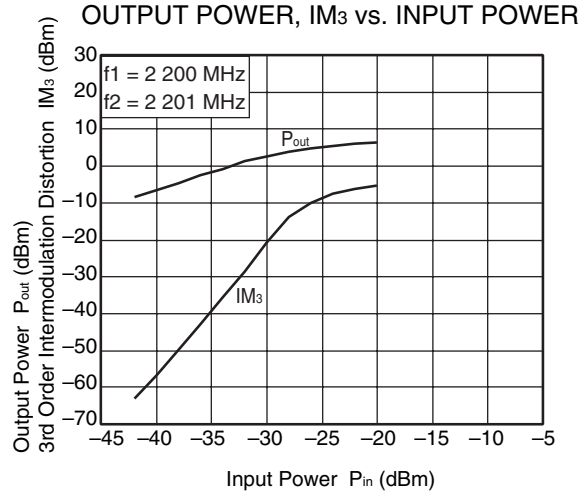
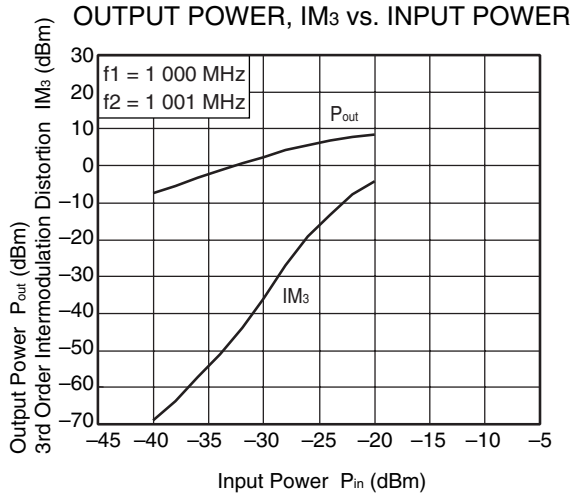
OUTPUT RETURN LOSS vs. FREQUENCY



Remark The graphs indicate nominal characteristics.



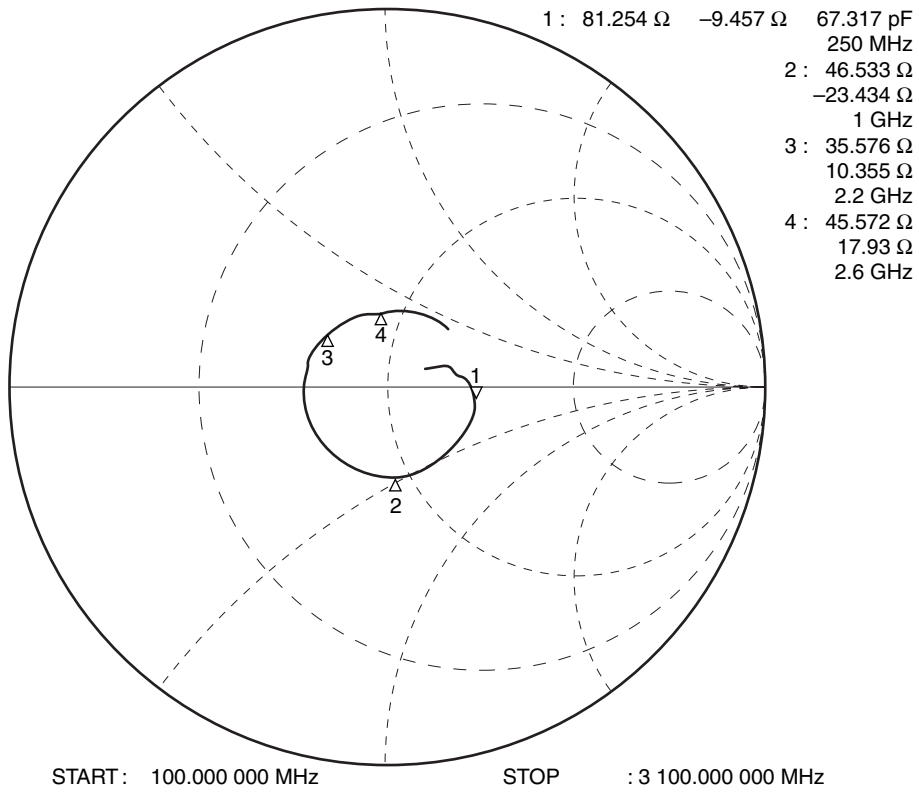
Remark The graphs indicate nominal characteristics.



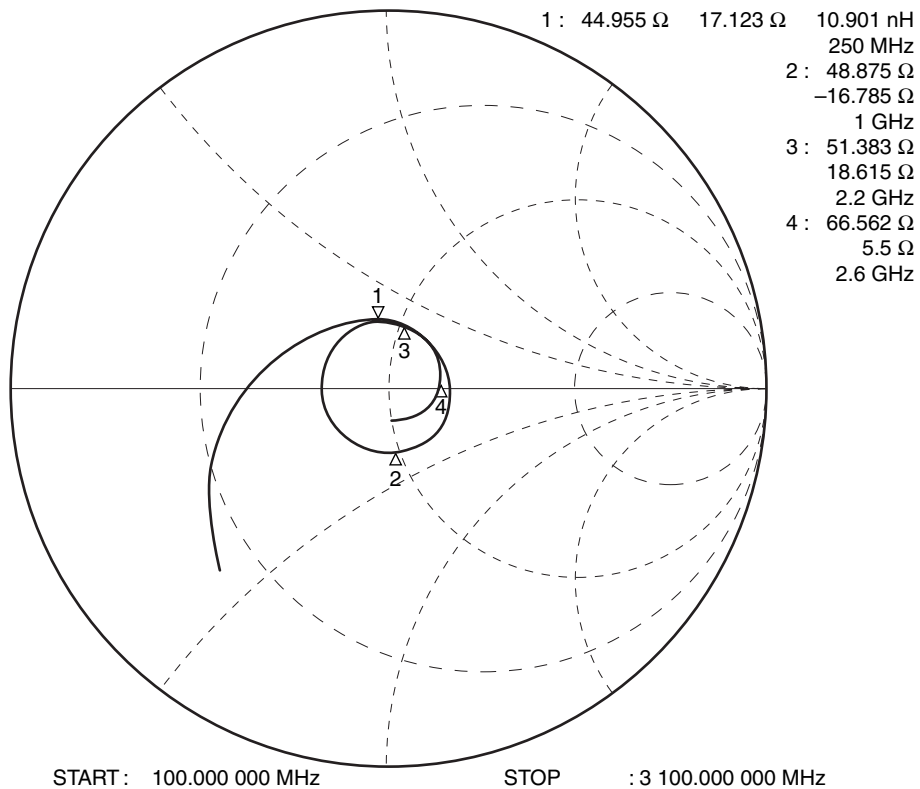
Remark The graphs indicate nominal characteristics.

S-PARAMETERS ($T_A = +25^\circ\text{C}$, $V_{DD} = V_{CC} = 5.0\text{ V}$, $P_{in} = -35\text{ dBm}$)

S₁₁-FREQUENCY



S₂₂-FREQUENCY



S-PARAMETERS

S-parameters/Noise parameters are provided on our web site in a form (S2P) that enables direct import to a microwave circuit simulator without keyboard input.

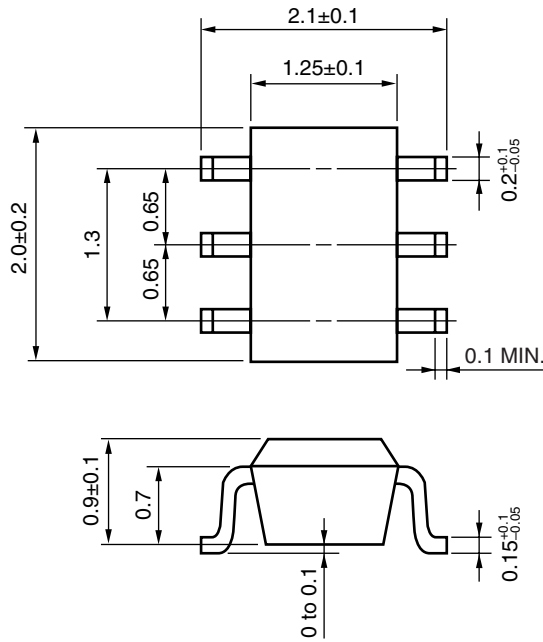
Click here to download S-parameters.

[RF and Microwave] → [Device Parameters]

URL <http://www.ncsd.necel.com/microwave/index.html>

PACKAGE DIMENSIONS

6-PIN SUPER MINIMOLD (UNIT: mm)



NOTES ON CORRECT USE

- (1) Observe precautions for handling because of electro-static sensitive devices.
- (2) Form a ground pattern as widely as possible to minimize ground impedance (to prevent undesired oscillation).
All the ground terminals must be connected together with wide ground pattern to decrease impedance difference.
- (3) The bypass capacitor should be attached to the V_{CC} line.
- (4) The inductor (L) must be attached between V_{CC} and output pins. The inductance value should be determined in accordance with desired frequency.
- (5) The DC cut capacitor must be attached to input and output pin.

RECOMMENDED SOLDERING CONDITIONS

This product should be soldered and mounted under the following recommended conditions. For soldering methods and conditions other than those recommended below, contact your nearby sales office.

| Soldering Method | Soldering Conditions | Condition Symbol |
|------------------|---|------------------|
| Infrared Reflow | Peak temperature (package surface temperature) : 260°C or below Time at peak temperature : 10 seconds or less Time at temperature of 220°C or higher : 60 seconds or less Preheating time at 120 to 180°C : 120±30 seconds Maximum number of reflow processes : 3 times Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below | IR260 |
| Wave Soldering | Peak temperature (molten solder temperature) : 260°C or below Time at peak temperature : 10 seconds or less Preheating temperature (package surface temperature) : 120°C or below Maximum number of flow processes : 1 time Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below | WS260 |
| Partial Heating | Peak temperature (terminal temperature) : 350°C or below Soldering time (per side of device) : 3 seconds or less Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below | HS350 |

Caution Do not use different soldering methods together (except for partial heating).

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